



**APPARATUS AND METHOD FOR TRANSMITTING/RECEIVING**  
**UPLINK DATA RETRANSMISSION REQUEST IN A CDMA**  
**COMMUNICATION SYSTEM**

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**PRIORITY**

This application claims priority under 35 U.S.C. § 119 to an application entitled "Apparatus and Method for Transmitting/Receiving Uplink Data Retransmission Request in a CDMA Communication System" filed in the 10 Korean Intellectual Property Office on January 4, 2003 and assigned Serial No. 2003-462, the contents of which are incorporated herein by reference.

**BACKGROUND OF THE INVENTION**

15        **1. Field of the Invention**

The present invention relates generally to a CDMA communication system, and in particular, to an apparatus and method for transmitting/receiving an uplink data retransmission request.

20        **2. Description of the Related Art**

In general, owing to the development of communication technology, CDMA (Code Division Multiple Access) has evolved into a system that enables high-speed packet data transmission. Such a communication system is commonly referred to as HSDPA (High-Speed Downlink Packet Access). HSDPA 25 generically refers to a data transmission scheme involving a high-speed downlink shared channel (HS-DSCH) supporting high-speed downlink packet transmission, and its related control channels in a UMTS (Universal Mobile Telecommunication System) developed in Europe. To support HSDPA, AMC (Adaptive Modulation and Coding), HARQ (Hybrid Automatic Retransmission 30 Request), and FCS (Fast Cell Select) were proposed. With reference to FIG. 1, the architecture of a WCDMA (Wideband Code Division Multiple Access) or UMTS communication system will be described below.

FIG. 1 is a block diagram illustrating the configuration of a typical 35 WCDMA communication system.

The WCDMA communication system comprises a core network (CN) 100, a plurality of RNSs (Radio Network Subsystems) 110 and 120, and a UE (User Equipment) 130. Each of the RNSs 110 and 120 includes an RNC (Radio 5 Network Controller) and a plurality of Node Bs (Node B and cell are used interchangeably, hereinafter). For example, the RNS 110 includes an RNC 111 and Node Bs 113 and 115, whereas the RNS 120 includes an RNC 112 and Node Bs 114 and 116. There are three types of RNCs, a serving RNC (SRNC), a drift RNC (DRNC) and a controlling RNC (CRNC) according to their functions. The 10 SRNC is distinguished from the DRNC according to their roles as relating to a UE. The SRNC is responsible for managing information related to the UE and data transmission between the UE and a CN (Core Network). If the UE transmits/receives data to/from the SRNC via an RNC which is not currently serving the UE, this RNC is the DRNC. An RNC controlling a Node B is a 15 CRNC for the Node B. In the case illustrated in FIG. 1, if the RNC 111 manages information related to UE 130, it is the SRNC for the UE 130. As the UE 130 moves and transmits/receives data through the RNC 112, the RNC 112 is the DRNC for the UE 130. The RNC 111, which controls the Node B 113, is the CRNC for the Node B 113.

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With reference to FIG. 1, HARQ, particularly n-channel SAW HARQ (n-channel Stop And Wait Hybrid Automatic Retransmission Request) will be described.

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With regard to the n-channel SAW HARQ, two new schemes were introduced to increase the efficiency of SAW ARQ (Stop And Wait Automatic Retransmission Request).

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One of the new schemes is soft combining. Soft combining is a scheme in which a receiver temporarily stores defective data and combines a retransmitted version of the data with the stored data in order to reduce error probability. There are two soft combining methods: chase combining (CC) and incremental redundancy (IR).

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In the CC method, a transmitter adopts the same format at both an initial

transmission and a retransmission. If  $m$  symbols are transmitted in one coded block at an initial transmission, the same  $m$  symbols are transmitted at a retransmission. The coded block is a unit of user data transmitted for one TTI (Transmit Time Interval). The same coding rate applies to both the initial 5 transmission and the retransmission. The receiver then combines the initially transmitted coded block with the retransmitted coded block, checks the CRC (Cyclic Redundancy Check) of the combined coded block, and determines if there are errors in the combined coded block.

10 The IR method uses different formats at an initial transmission and a retransmission. If  $m$  symbols are generated from  $n$ -bit user data after channel coding, the transmitter transmits part of the  $m$  symbols at an initial transmission and sequentially transmits the remaining symbols at a retransmission. Different coding rates apply to the initial transmission and the retransmission. The receiver 15 then produces a coded block with a high coding rate by attaching the retransmitted coded block to the initially transmitted coded block, and corrects errors in the coded block. In the IR scheme, an initial transmission and retransmissions are identified by their version numbers. The initial transmission is numbered 1, a first retransmission is numbered 2, and the following 20 retransmission is numbered 3. By using this version information, the receiver can correctly combine the initially coded block with any retransmitted coded blocks.

The IR method is implemented in either a self-decodable or a non-self-decodable format. Self-decodable and partial IR are interchangeably used, 25 whereas non-self-decodable and full IR are interchangeably used. Hereinafter, the terms, partial IR and full IR will mainly be used. The partial IR uses a part of an initial transmission format at a retransmission. This part of the initial transmission format is the systematic part of a turbo code. The systematic part enables self-decoding. If the partial IR is adopted, the receiver can decode 30 received data without combining buffered initially transmitted data with retransmitted data. On the other hand, the full IR uses different formats at an initial transmission and a retransmission, to thereby maximize redundancy information-incurred gain. Because a systematic part is not included in retransmitted data in the full IR, it is impossible to decode received data with 35 retransmitted data. Therefore, the receiver can decode the received data normally

only if it combines the initially transmitted data with the retransmitted data.

The other scheme of increasing the efficiency of the n-channel SAW HARQ is HARQ.

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In a general SAW HARQ, a Node B transmits the current packet only when it receives an acknowledgement (ACK) of the receipt of the previously transmitted packet. Thus, it may occur that even when the Node B can transmit the current packet, the Node B must wait for the ACK. The n-channel SAW HARQ allows transmission of successive packets without receiving an ACK about a previously transmitted packet, thereby increasing the use efficiency of a radio link. In the n-channel SAW HARQ, n logical channels are established between a Node B and a UE. The UE determines upon which channel a packet received at a particular time point is mapped by identifying the n logical channels 10 by predetermined time points or explicit channel numbers assigned to them. The UE must rearrange packets in the correct order, or soft-combine the packets. The n-channel SAW HARQ will be described in more detail referring to FIG. 1. It is assumed herein that the n-channel SAW HARQ, particularly a 4-channel SAW HARQ is implemented between the UE 130 and the Node B 114 and logical IDs 15 1 to 4 are assigned to the four channels. The UE 130 and the Node B 114 are each provided with an HARQ processor for each channel in the physical layer. The Node B 114 assigns channel ID 1 to an initial transmission coded block and transmits the coded block to the UE 130. The channel ID can be explicitly assigned, or implicitly assigned as a predetermined time point. If the coded block 20 25 with channel ID 1 has errors, the UE 130 provides the coded block to an HARQ processor for channel ID 1, namely HARQ processor 1, and transmits a non- acknowledgement (NACK) about channel 1 to the Node B 114. The Node B 114 can transmit the next coded block on channel 2 irrespective of whether it has received an ACK about the coded block on channel 1. If the next coded block 30 35 also has errors, the Node B 114 also transmits the next coded block to a corresponding HARQ processor. Upon receipt of the NACK about the coded block on channel 1 from the UE 130, the Node B retransmits the coded block on channel 1. The UE 130 recognizes that the received coded block is a retransmitted version of the previous coded block received on channel 1 and transmits the retransmitted coded block to HARQ processor 1. HARQ processor

1 soft-combines the initially transmitted coded block with the retransmitted coded block. As described above, the n-channel SAW HARQ matches a channel ID to an HARQ processor on a one-to-one correspondence. Without delaying transmission of user data until an ACK is received, an initial transmission and  
5 retransmissions can be appropriately matched.

As described above, the process of determining if received data has errors and correspondingly transmitting an ACK/NACK in the receiver is quite significant to efficiently support the HARQ scheme. The transmitter determines  
10 whether to retransmit the data according to the ACK/ NACK. In HSDPA, an uplink HS-DPCCH (High Speed-Dedicated Physical Control Channel) delivers an ACK/NACK about data transmitted by a transmitter or a Node B. With respect to the HS-DPCCH, if an uplink control channel slot format used for a non-HSDPA communication system, for example, Release-99, is modified to  
15 deliver an ACK/NACK, compatibility with the Release-99 communication system is not ensured and an uplink channel structure becomes complex. Thus, the HS-DPCCH is defined using a novel channelization code.

Control information delivered on the HS-DPCCH includes ACK/NACK  
20 and CQI (Channel Quality Indicator). The ACK/NACK can be expressed in one bit. As to the CQI, upon receipt of a downlink channel signal, a UE measures channel quality from the downlink channel signal and transmits a CQI representing the channel quality to a Node B. The Node B determines an MCS (Modulation and Coding Scheme) level for the HS-DSCH according to the  
25 channel quality and generates a TFRI (Transport Format and Resource Related Information) as control information about the HS-DSCH. For example, if the CQI indicates a good channel condition, the Node B selects a modulation that exhibits a high BER (Bit Error Rate) but allows a high data rate, such as 16-QAM (Quadrature Amplitude Modulation). On the contrary, if the CQI indicates  
30 a poor channel condition, the Node B selects a relatively reliable modulation such as QPSK (Quadrature Phase Shift Keying). The ACK/NACK and CQI are delivered over the HS-DPCCH. If the HS-DPCCH has a 3-slot TTI structure, the ACK/NACK is delivered in one of the three slots and the CQI in the remaining two slots.

Studies have been actively conducted on uplink communication systems like the HSDPA communication system for improving uplink communication efficiency. Currently an uplink communication system which enables uplink data transmission on an enhanced uplink dedicated channel (EUDCH) is being 5 proposed. This EUDCH communication system still uses the schemes adopted in the HSDPA communication system. It adapts to AMC and HARQ. Also, the EUDCH communication system can use a short TTI of 2ms (3 slots) similar to the HSDPA communication system. The TTI is a unit time period for which one coded block is transmitted. Downlink channel scheduling is carried out in a Node 10 B, to thereby prevent scheduling-caused delay.

The EUDCH communication system transmits data on the uplink and needs HARQ for transmitted uplink data, as described above in the context of the HSDPA communication system. To support HARQ, the process of transmitting 15 an ACK/NACK from a receiver to a transmitter is essential.

## **SUMMARY OF THE INVENTION**

An object of the present invention is to provide an apparatus and method 20 for requesting uplink data retransmission in a CDMA communication system.

Another object of the present invention is to provide an apparatus and method for requesting uplink data retransmission by puncturing a data field of a DL DPCH in a CDMA communication system.

25 A further object of the present invention is to provide an apparatus and method for randomly determining a position to puncture in a data field of a DL DPCH in order to insert an uplink data retransmission request in the punctured position in a CDMA communication system.

30 Still another object of the present invention is to provide an apparatus and method for requesting uplink data transmission taking into consideration the compatibility with other systems in a CDMA communication system.

35 The above objects are achieved by providing an apparatus and method

for transmitting/receiving an uplink data retransmission request.

According to one aspect of the present invention, in an apparatus for requesting uplink data retransmission in a CDMA communication system using a DL DPCH to which a DL DPCCH and a DL DPDCH are mapped, the DL DPCCH including a TPC field, a TFCI field, and a pilot field, and the DL DPDCH including first and second data fields for delivering downlink data, a puncturer generates a p-bit ACK or a p-bit NACK according to whether data received on an EUDCH is normal or abnormal, and punctures p bits in a position 5 to transmit the ACK or NACK at in the first and second data fields of the DL DPDCH, determined under a predetermined control. A puncturing controller determines the position to transmit the ACK or NACK in the first and second data fields of the DL DPDCH. A DL DPCH transmitter inserts the ACK or NACK in the punctured bit positions and transmits the DL DPCH with the ACK 10 or NACK.

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According to another aspect of the present invention, in an apparatus for requesting uplink data retransmission in a CDMA communication system using a DL DPCH to which a DL DPCCH and a DL DPDCH are mapped, the DL DPCCH including a TPC field, a TFCI field, and a pilot field, and the DL DPDCH including first and second data fields for delivering downlink data, a DL DPCH receiver transmits data on an EUDCH and receives the DL DPCH signal. A puncturing controller determines a position to receive a p-bit ACK or a p-bit NACK in the first and second data fields of the DL DPDCH. A puncturer 20 25 extracts p bits at the decided position as the ACK or NACK.

According to a further aspect of the present invention, in a method for requesting uplink data retransmission in a CDMA communication system using a DL DPCH to which a DL DPCCH and a DL DPDCH are mapped, the DL DPCCH including a TPC field, a TFCI field, and a pilot field, and the DL DPDCH including first and second data fields for delivering downlink data, data 30 is received on an EUDCH, a p-bit ACK is generated if the received data is normal, and a p-bit NACK is generated if the received data is abnormal. A position to transmit the ACK or NACK is determined in the first and second data fields of the DL DPDCH. p bits are punctured in the decided position, the ACK 35

or NACK is inserted in the punctured bit positions, and the DL DPCH with the ACK or NACK is transmitted.

According to still another aspect of the present invention, in a method for  
5 requesting uplink data retransmission in a CDMA communication system using a  
DL DPCH to which a DL DPCCH and a DL DPDCH are mapped, the DL  
DPCCH including a TPC field, a TFCI field, and a pilot field, and the DL  
DPDCH including first and second data fields for delivering downlink data, data  
is transmitted on an EUDCH, and the DL DPCH signal is received. A position to  
10 receive a p-bit ACK or a p-bit NACK is determined in the first and second data  
fields of the DL DPDCH. p bits are extracted from the decided position as the  
ACK or NACK.

According to yet another aspect of the present invention, in a method for  
15 requesting uplink data retransmission in a CDMA communication system using a  
downlink dedicated data channel for delivering downlink data, data is received  
on an uplink dedicated channel, a p-bit ACK is generated if the received data is  
normal, and a p-bit NACK is generated if the received data is abnormal. A position to transmit the ACK or NACK is determined in the downlink dedicated  
20 data channel. p bits are punctured in the decided position, the ACK or NACK is  
inserted in the punctured bit positions, and the downlink dedicated data channel  
with the ACK or NACK is transmitted.

#### **BRIEF DESCRIPTION OF THE DRAWINGS**

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FIG. 1 is a block diagram illustrating the configuration of a typical  
WCDMA communication system;

FIG. 2 is a diagram illustrating a signal flow for a data retransmission in  
an EUDCH communication system;

30 FIG. 3 is a block diagram illustrating the structure of a DL DPCH in the  
typical WCDMA communication system;

FIG. 4 is a block diagram illustrating the structure of a DL DPCH that  
delivers an ACK/NACK relating to uplink data according to an embodiment of  
the present invention;

35 FIG. 5 is a block diagram illustrating the structure of a DL DPCH that

delivers an ACK/NACK relating to uplink data according to another embodiment of the present invention;

FIG. 6 is a block diagram of a Node B transmitter supporting the DL DPCH structure illustrated in FIG. 4;

5 FIG. 7 is a block diagram of a Node B transmitter supporting the DL DPCH structure illustrated in FIG. 5;

FIG. 8 is a block diagram of a UE receiver corresponding to the Node B transmitter illustrated in FIG. 6;

10 FIG. 9 is a block diagram of a UE receiver corresponding to the Node B transmitter illustrated in FIG. 7; and

FIG. 10 is a flowchart illustrating an operation for transmitting an ACK/NACK about uplink data according to the embodiments of the present invention.

## 15 **DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS**

Preferred embodiments of the present invention will be described herein below with reference to the accompanying drawings. In the following description, well-known functions or constructions are not described in detail since they 20 would obscure the invention in unnecessary detail.

FIG. 2 is a diagram illustrating a signal flow for a data retransmission process in an EUDCH communication system.

25 The EUDCH communication system is being studied to determine methods to increase uplink communication efficiency. Uplink data transmission is carried out on an uplink channel, EUDCH. The EUDCH communication system can still use the schemes as adopted for the HSDPA communication system, as described before, i.e. It can use AMC and HARQ schemes.

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Referring to FIG. 2, the EUDCH is set up between a Node B 201 and a UE 202 in step 203. The EUDCH setup is carried out by message transmission/reception on dedicated transport channels. In step 204, the UE 202 reports the channel condition of the EUDCH to the Node B 201 in step 204 35 (Channel Report). The channel condition can be represented by the EUDCH

transmit power. The Node B 201 estimates the uplink channel condition of the UE 202 based on the reported channel condition information. If the channel condition information is the EUDCH transmit power, the Node B 201 can estimate the reception power of the EUDCH at the Node B 201 from the EUDCH 5 transmit power. Thus, the Node B 201 estimates the current channel condition by comparing the EUDCH transmit power reported by the UE 202 to the reception power of the EUDCH measured at the Node B 201.

In step 205, the Node B 201 performs scheduling based on the estimated 10 channel condition of the UE 202 and transmits the scheduling result to the UE 202 (Rate Indication). The scheduling refers to the process for selecting a UE to transmit packet data for the next TTI among a plurality of UEs communicating on the EUDCH within the same cell and determining a modulation scheme for the packet data, the number of codes to be assigned to the data transmission, and 15 the data rate. In FIG. 2, the scheduling result indicates the data rate, by way of example. The UE 202 receives the scheduling result from the Node B 201 and transmits packet data based on the scheduling result. That is, the UE 202 generates a TFRI from the scheduling result and transmits the TFRI to the Node B 201 in step 206. The TFRI includes an orthogonal variable spreading factor 20 (OVSF) code applied to the EUDCH, a modulation scheme, a data size, and HARQ information.

After transmitting the TFRI, the UE 202 determines the data rate of the 25 packet data to be transmitted based on the TFRI and transmits the packet data at the determined data rate over the EUDCH to the Node B 201 in step 207 (UL Packet Data Transmission). The Node B 201 determines if the received packet data is normal. If the packet data is normal, the Node B 201 transmits an ACK to the UE 202. If the received data is abnormal, the Node B 201 transmits an NACK to the UE 202 in step 208. In the case of the ACK, the UE transmits the next 30 packet data, while in the case of the NACK, the UE retransmits the transmitted packet data in step 209 (New Data or Retransmission). In either case, steps 204, 205 and 206 are repeated. As described before, the format of the retransmitted packet data is different depending on the soft combining scheme used to support the HARQ. If the EUDCH communication system employs the CC method, the 35 initially transmitted packet data and the retransmitted packet data are in the same

format. If the soft combining is the IR method, the initially transmitted packet data and the retransmitted packet data are in different formats. If the IR is self-decodable, namely partial IR, the initial transmission format is partially identical to the retransmission format. If the IR is non-self-decodable, namely full IR, the 5 initial transmission format is entirely different from the retransmission format.

The Node B requests a retransmission of received uplink data, taking into consideration a channel condition to deliver the retransmitted data. The present invention proposes a method of transmitting an ACK/NACK related to 10 the uplink data.

First, a novel downlink shared control channel can be defined as a channel to deliver the ACK/NACK.

15 In view of the nature of a shared channel, the use of the downlink shared control channel limits the number of UEs that can concurrently access the channel.

Secondly, a novel downlink dedicated control channel can be defined as 20 a channel to deliver the ACK/NACK.

Compared to the downlink shared control channel, the downlink dedicated channel does not limit the number of UEs that can simultaneously access the channel. Despite this advantage, the use of the downlink dedicated 25 channel may cause problems regarding compatibility with existing systems.

Thirdly, an existing downlink dedicated channel can be defined as a channel to deliver the ACK/NACK.

30 This method causes less problems regarding compatibility with the existing systems and does not limit the number of UEs that can simultaneously access the channel, which is encountered with the use of the downlink shared control channel.

35 The present invention provides the method of transmitting an

ACK/NACK on the existing downlink dedicated channel.

The structure of a DL DPCH (Downlink Dedicated Physical Control Channel) in the current WCDMA communication system will be described with  
5 reference to FIG. 3.

FIG. 3 is a diagram illustrating the structure of the DL DPCH in the typical WCDMA communication system.

10 Referring to FIG. 3, each DL DPCH frame includes 15 slots, slot#0 to slot#14. Each of the slots comprises a DPDCH (Dedicated Physical Data Channel) for delivering upper-layer data from a Node B to a UE, and a DPCCH (Dedicated Physical Control Channel) for transmitting a physical layer control signal. The DPCCH has a TPC (Transport Power Control) 302, a TFCI 15 (Transport Format Combination Indicator) 303, and pilot bits 305. As illustrated in FIG. 3, each slot in one DL DPCH frame is 2,560 chips in length. Data 1 301 and Data 2 304 represent upper-layer data transmitted from the Node B to the UE on the DPDCH. The TPC 302 provides information for controlling the transmit power of the UE. The TFCI 303 indicates a TFC (Transport Format Combination) adopted by the downlink channel in the current frame (10ms). Lastly, the pilot bits 305 provides a criterion by which the UE controls the transmit power of a DPCH. The information in the TFCI 303 is divided into a dynamic part and a semi-static part. There are TBS (Transport Block Size) and TBSS (Transport Block Set Size) in the dynamic part. The semi-static part 20 25 provides information about TTI, channel coding scheme, coding rate, static rate matching, CRC size. Thus, the TFCI 303 indicates the number of transport blocks (TBs) on one channel frame and the number of TFCs available to each of the TBs.

30 With reference to FIG. 4, a description will be made of the structure of a DL DPCH for transmitting an ACK/NACK related to the uplink data according to an embodiment of the present invention.

FIG. 4 is a diagram illustrating the structure of a DL DPCH for 35 transmitting an ACK/NACK related to the uplink data according to an

embodiment of the present invention.

As described earlier, an ACK/NACK related to the uplink data must be transmitted to support HARQ in the EUDCH communication system. While the 5 structure of an existing downlink dedicated channel is still used, predetermined bits of the DPDCH in the DL DPCH are punctured to transmit the ACK/NACK in the present invention.

As illustrated in FIG. 4, the DL DPCH comprises the DPDCH and the 10 DPCCH. The DPDCH has Data 1 401 and Data 2 404, while the DPCCH has a TPC 402, a TFCI 403, and pilot bits 405. Data 1 401 and Data 2 404 are identical to Data 1 301 and Data 2 304 as illustrated in FIG. 3. The TPC 402, TFCI 403, and pilot 405 are identical to the TPC 302, TFCI 303 and pilot bits 305 as 15 illustrated in FIG. 3, respectively. One thing to note herein is that the predetermined bits of a data field, for example,  $p$  bits of Data 2 404, are punctured and an ACK/NACK 406 related to the uplink data is inserted in the punctured  $p$  bits. The  $p$ -bit puncturing does not substantially affect the performance of data transmission on the DPDCH. However, if the punctured  $p$ -bit positions are fixed, the puncturing may deteriorate the data transmission 20 performance. Thus, the positions of the  $p$  bits are randomly selected.

The bit positions of the DPDCH to be punctured to transmit the ACK/NACK are determined as in Equation 1 by

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$$P(i) = \text{rand}(N_{\text{data}} - p + 1)$$

..... (1)

where  $P(i)$  indicates the first bit position to be punctured in an  $i$ th slot,  $\text{rand}(x)$  is a function for generating a random variable in a range from 0 to  $x-1$ ,  $N_{\text{data}}$  is the number of data bit positions in one DL DPCH slot, and  $p$  is the number of bits 30 required to transmit the ACK/NACK. As noted from Equation (1), the ACK/NACK is transmitted in  $p$  successive bits randomly selected from a data field of one DL DPCH slot. That is, the bits of Data 1 401 and Data 2 404 in one DL DPCH slot are arranged together and sequentially numbered, starting with 0 for the first bit. Then  $p$  successive bits from the position calculated by Equation 35 (1) are punctured and the ACK/NACK is transmitted in the punctured bit

positions. Although the ACK/NACK can be represented in one bit, it occurs p times in each slot, that is, it is transmitted in p bits so as to increase radio transmission reliability. On the assumption that one TTI has N slots in the EUDCH communication system, p-bit ACK/NACK information can be 5 transmitted in  $\frac{p}{N}$  bits per slot for N slots, or fully transmitted in one slot preset between the Node B and the UE in one TTI.

The case of repeating the ACK/NACK N times in each slot will be described with reference to FIG. 5.

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FIG. 5 is a diagram illustrating the structure of a DL DPCH that delivers an ACK/NACK related to the uplink data according to another embodiment of the present invention.

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An ACK/NACK transmission period is based on a scheduling period. A Node B transmits an ACK/NACK at least once within the scheduling period. Equation (1) applies to the case where an ACK/NACK is transmitted in each slot, whereas Equation (2) applies to the case where a p-bit ACK/NACK is transmitted through all of the slots of a TTI. In this case successive  $\lfloor p/N \rfloor$  bits 20 are punctured and corresponding ACK/NACK is transmitted in each slot. And then for the last slot in the TTI, the remaining ACK/NACK is transmitted.

$$P(i) = \begin{cases} \text{rand}(N_{\text{data}} - \lfloor p/N \rfloor + 1), & n = 0, 1, \dots, N-2 \\ \text{rand}(N_{\text{data}} - (p - \lfloor p/N \rfloor \times (N-1)) + 1), & n = N-1 \end{cases} \dots \dots \dots \quad (2)$$

25 where P(i) indicates the first bit position to be punctured in an ith slot,  $\lfloor x \rfloor$  is a maximum natural number equal to or less than x, rand(x) is a function for generating a random variable in a range from 0 to x-1,  $N_{\text{data}}$  is the number of data bits in one DL DPCH slot, p is the number of bits required to transmit the ACK/NACK, n is a slot index in a TTI (0, 1, ..., N-1), and N is the number of 30 slots in one TTI. Here, n=i modulo N. Modulo is the remainder of a division. Uniformly distributed transmission of the ACK/NACK across all slots of a TTI according to Equation (2) improves transmission reliability.

The DL DPCH illustrated in FIG. 5 is configured to transmit the ACK/NACK in  $p/N$  bits per slot for  $N$  slots according to Equation (2) under the assumption that one TTI has  $N$  slots. For example, if the Node B schedules transmission based on a 3-slot TTI of 2ms in the EUDCH communication system,

5 the ACK/NACK must be transmitted at least once for each 2-ms TTI. Relying on Equation (1), the ACK/NACK is transmitted in  $p$  bits in each slot. Therefore, the ACK/NACK is 3 bits in total within one TTI. If the UE and the Node B agree that the ACK/NACK is to be transmitted in the first slot, the  $p$ -bit ACK/NACK is obviously transmitted in one TTI. On the other hand, if Equation (2) is used, the

10  $p$ -bit ACK/NACK is separately transmitted in  $p/3$  bits per slot for the three slots of a TTI. Referring to Equation (1) and Equation (2), the ACK/NACK transmission can be correctly performed if the positions of the ACK/NACK are preset between the Node B and the UE.

15 Even though the UE transmits packet data on the EUDCH, the Node B may fail to receive the packet data. In this case, the Node B does not transmit an ACK/NACK on the DL DPCH. The Node B leaves the data of the DL DPCH unpunctured. The UE, however, awaits the ACK/NACK for the transmitted packet data and extracts actual data as the ACK/NACK, causing errors. To

20 prevent these errors, the Node B punctures predetermined bits of the DL DPCH in DTX (Discontinuous Transmission) despite non-reception of packet data on the EUDCH in accordance with the present invention.

Equation (1) and Equation (2) have defined the rules of transmitting an

25 ACK/NACK. Next, a detailed description will be made of how the Node B actually puncture  $P$  bit positions to transmit the ACK/NACK with reference to Equation (3) and Equation (4).

In general, Node Bs are asynchronous with each other in the WCDMA

30 communication system. Hence, no timing synchronization is provided between them. Each Node B has its own timer and operates based on a reference timing counted by the timer. The timer counts in units of BFN (Node B Frame Number). Each Node B may cover a plurality of cells and each of the cells is provided with a timer operating with a predetermined offset from the BFN. The timer in the cell

35 counts in units of SFN (System Frame Number). One SFN is 10ms in duration

and numbered between 0 and 4095. One SFN includes 38,400 chips. Hence, one chip is 10ms/38,400 in duration. Using the SFN, each cell transmits an ACK/NACK in a different position from other cells within a data field of the DL DPCH, which can be expressed as Equation 3:

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$$P(i) = \{SFN \times 15\text{slots} + \text{current\_slot\_number}\} \bmod (N_{\text{data}} - p + 1) \quad \dots \dots (3)$$

where  $P(i)$  is the first bit position to be punctured in an  $i$ th slot,  $\bmod$  represents the modulo operation,  $\text{current\_slot\_number}$  is the current slot index, SFN is the 10 SFN of the current cell,  $N_{\text{data}}$  is the number of data bits in one DL DPCH slot, and  $p$  is the number of bits required to transmit the ACK/NACK.

$\{SFN \times 15\text{slots} + \text{current\_slot\_number}\}$  in Equation (3) is the SFN of the current cell expressed in terms of slots. The first position to insert the 15 ACK/NACK in a field of the DL DPCH in the current slot is randomly decided by modulo-operating  $\{SFN \times 15\text{slots} + \text{current\_slot\_number}\}$  with  $(N_{\text{data}} - p + 1)$ . The current slot index is known by counting the number of slots in the state where the UE acquires frame synchronization. The SFN can be replaced by CFN (Connection Frame Number). The CFN corresponds to a DPCH frame 20 number, ranging from 0 to 255.

In the meantime, the ACK/NACK can be transmitted by being distributed across the slots of a TTI, as described earlier in connection with Equation (2). Then, Equation (3) is developed to Equation (4):

25

$$\begin{aligned} P(i) &= \{SFN \times 15\text{slots} + \text{current\_slot\_number}\} \bmod (N_{\text{data}} - \lfloor p/N \rfloor + 1), \quad n = 0, 1, \dots, N-2 \\ P(i) &= \{SFN \times 15\text{slots} + \text{current\_slot\_number}\} \bmod (N_{\text{data}} - (p - \lfloor p/N \rfloor \times (N-1)) + 1), \\ n &= N-1 \end{aligned} \quad \dots \dots (4)$$

where  $p$  is the number of bits required to transmit the ACK/NACK,  $N_{\text{data}}$  is the 30 number of data bits in one DL DPCH slot,  $n$  is a slot index in a TTI ( $n=0, 1, \dots, N-1$ ), and  $N$  is the number of slots in the TTI. Here,  $n=i \bmod N$ . The CFN can be used instead of the SFN, as described in connection with Eq. (3).

The SFN in Equation (3) and Equation (4) is different for each cell.

Therefore, if the UE transmits uplink data on the same EUDCH in a soft handover zone, each cell places an ACK/NACK about the uplink data in a different position. As a result, the UE achieves diversity gain. As far as 'a' is an integer multiple of 'b' in an operation of 'a mod b',  $P(i)$  can be the same for each 5 cell. This can be prevented by substituting the CFN for the SFN in Equation (3) and Equation (4) and assigning a different offset to each cell, thereby allowing each cell to position the ACK/NACK differently.

Now, the structure of a Node B transmitter according to the first 10 embodiment of the present invention will be described with reference to FIG. 6.

FIG. 6 is a block diagram of a Node B transmitter supporting the DL DPCCH structure illustrated in FIG. 4.

15 The illustrated Node B transmitter is configured to correspond with the DL DPCCH that delivers a 1-bit ACK/NACK  $p$  times in one slot as illustrated in FIG. 4. For conciseness, only the DL DPCCH will be considered in the Node B transmitter structure.

20 Referring to FIG. 6, a puncturing controller 606 in the Node B determines the positions to be punctured in the DL DPCCH through an initial setup with a UE so that an ACK/NACK related to the uplink data received on the EUDCH from the UE can be inserted in the punctured positions. The puncturing controller 606 randomly determines the puncturing positions as described in 25 connection with Equation (1) and Equation (3). Upon receipt of uplink data on the EUDCH from the UE, the Node B determines if the uplink data is normal and generates an ACK/NACK according to the determination. The ACK/NACK is represented in one bit and occurs  $p$  times to improve its transmission reliability. A repeater 604 repeats the 1-bit ACK/NACK to  $p$  bits and outputs the repeated 30 ACK/NACK to a puncturer 607. A DL DPCH signal to be transmitted is also applied to the puncturer 607.

The puncturer 607 punctures the corresponding  $p$  bits in a data field of the DL DPCCH under the control of the puncturing controller 606 and inserts the 35 ACK/NACK received from the repeater 604 in the punctured  $p$  bit positions. A

serial to parallel converter (SPC) 608 converts the signal received from the puncturer 607 to I and Q bit streams and outputs the bit streams to a spreader 609. The spreader 609 includes multipliers 621 and 623. The multiplier 621 multiplies the I bit stream by a spreading code  $C_{OVSF}$ , and the multiplier 623 multiplies the 5 Q bit stream by the spreading code  $C_{OVSF}$ . The outputs of the multipliers 621 and 623 are fed to a summer 611 and a multiplier 610, respectively. The multiplier 610 converts the signal received from the multiplier 623 to an imaginary number component by multiplying the signal by a component  $j$ . The summer 611 sums the outputs of the multipliers 621 and 610 to a chip rate level complex signal. A 10 multiplier 612, serving as a scrambler, multiplies the output of the summer 611 by a scrambling code  $C_{SCRAMBLE}$ . A multiplier 613 multiplies the scrambled signal by a predetermined channel gain. A modulator 614 modulates the output of the multiplier 613 in a predetermined modulation scheme. An RF processor 615 converts the modulated signal to an RF signal and transmits the RF signal in 15 the air via an antenna 616.

With reference to FIG. 7, the structure of a Node B transmitter according to the second embodiment of the present invention will be described.

20 FIG. 7 is a block diagram of a Node B transmitter supporting the DL DPCCH structure illustrated in FIG. 5.

The illustrated Node B transmitter is configured to correspond to the DL DPCCH that delivers an ACK/NACK N times across the slots of one TTI as 25 illustrated in FIG. 5. For conciseness, only the DL DPCCH will be considered in the Node B transmitter structure.

Referring to FIG. 7, a puncturing controller 706 in the Node B determines the positions to be punctured in the DL DPCCH through an initial setup 30 with a UE so that an ACK/NACK related to the uplink data received on the EUDCH from a UE can be inserted in the punctured positions. The puncturing controller 706 randomly determines the puncturing positions as described in connection with Equation (2) and Equation (4). Upon receipt of uplink data on the EUDCH from the UE, the Node B determines if the uplink data is normal and 35 generates an ACK/NACK according to the determination. The ACK/NACK is

represented in one bit and repeated to p bits to improve its transmission reliability. A repeater 704 repeats the 1-bit ACK/NACK to p bits and outputs the repeated ACK/NACK to a buffer 705. The p-bit ACK/NACK is buffered because it is transmitted not in one slot at one time but distributedly in p/N bits per slot for N 5 slots of a TTI (on the assumption that one TTI has N slots). Under the control of the puncturing controller 706, p/N bits of the p-bit ACK/NACK per slot are fed to a puncturer 707 at bit positions where the ACK/NACK is to be transmitted. A DL DPCH signal to be transmitted is also applied to the puncturer 707.

10 The puncturer 707 punctures the corresponding p/N bits in a data field of the DL DPCH under the control of the puncturing controller 706 and inserts the ACK/NACK received from the buffer 705 in the punctured p/N bit positions. An SPC 708 converts the signal received from the puncturer 707 to I and Q bit streams and outputs the bit streams to a spreader 709. The spreader 709 includes 15 multipliers 721 and 723. The multiplier 721 multiplies the I bit stream by a spreading code  $C_{OVSF}$ , and the multiplier 723 multiplies the Q bit stream by the spreading code  $C_{OVSF}$ . The outputs of the multipliers 721 and 723 are fed to a summer 711 and a multiplier 610, respectively. The multiplier 710 converts the signal received from the multiplier 723 to an imaginary number component by 20 multiplying the signal by a component  $j$ . The summer 711 sums the outputs of the multipliers 721 and 710 to a chip rate level complex signal. A multiplier 712, serving as a scrambler, multiplies the output of the summer 611 by a scrambling code  $C_{SCRAMBLE}$ . A multiplier 713 multiplies the scrambled signal by a predetermined channel gain. A modulator 714 modulates the output of the 25 multiplier 713 in a predetermined modulation scheme. An RF processor 715 converts the modulated signal to an RF signal and transmits the RF signal in the air via an antenna 716.

30 The structure of a UE receiver according to the first embodiment of the present invention will be described with reference to FIG. 8.

FIG. 8 is a block diagram of a UE receiver that corresponds to the Node B transmitter illustrated in FIG. 6.

35 The illustrated UE receiver is configured to support the DL DPCH

illustrated in FIG. 4 which delivers an ACK/NACK  $p$  times in one slot. Notably, the UE receiver structure as illustrated focuses only on the DL DPCH for conciseness.

5 Referring to FIG. 8, a signal received from the air via an antenna 816 is fed to an RF processor 815. The RF processor 815 downconverts the received signal to a baseband signal. A demodulator 814 demodulates the baseband signal in a demodulation scheme corresponding to the modulation scheme adopted in the Node B transmitter. A multiplier 812, functioning as a descrambler, 10 multiplies the demodulated signal by a predetermined scrambling code,  $C_{SCRAMBLE}$ . An SPC 811 converts the descrambled signal to parallel I and Q bit streams. A despreader 809 has multipliers 821 and 823. The multiplier 821 multiplies the I bit stream by a spreading code  $C_{OVSF}$ , and the multiplier 823 multiplies the product of the Q bit stream and a  $j$  component, received from a 15 multiplier 810, by the spreading code  $C_{OVSF}$ . A channel compensator 805 channel-compensates the spread signals received from the multipliers 821 and 823. A summer 808 sums the channel-compensated I and Q bit streams and feeds the sum to a puncturer 807.

20 Meanwhile, a puncturing controller 806 in the UE determines the positions inserted with an ACK/NACK relating to the uplink data transmitted on the EUDCH through an initial setup with the Node B. The puncturing controller 806 determines the randomly inserted positions as described in connection with Equation (1) and Equation (3). The puncturer 807 extracts the ACK/NACK from 25 the inserted positions in the signal received from the summer 808, feeds the ACK/NACK to an ACK/NACK extractor 804, and outputs the remaining signal as a DL DPCH signal, under the control of the puncturing controller 806. The ACK/NACK extractor 804 converts the  $p$ -bit ACK/NACK to a 1-bit ACK/NACK.

30

The structure of a UE receiver according to the second embodiment of the present invention will be described with reference to FIG. 9.

FIG. 9 is a block diagram of a UE receiver that corresponds to the Node 35 B transmitter illustrated in FIG. 7.

The illustrated UE receiver is configured to support the DL DPCH illustrated in FIG. 5 which delivers an ACK/NACK N times across the slots of a TTI. Notably, the UE receiver structure as illustrated focuses only on the DL 5 DPCH for conciseness.

Referring to FIG. 9, a signal received from the air via an antenna 916 is fed to an RF processor 915. The RF processor 915 downconverts the received signal to a baseband signal. A demodulator 914 demodulates the baseband signal 10 in a demodulation scheme corresponding to the modulation scheme adopted in the Node B transmitter. A multiplier 912, functioning as a descrambler, multiplies the demodulated signal by a predetermined scrambling code,  $C_{SCRAMBLE}$ . An SPC 911 converts the descrambled signal to parallel I and Q bit streams. A despread 909 has multipliers 921 and 923. The multiplier 921 15 multiplies the I bit stream by a spreading code  $C_{OVSF}$ , and the multiplier 923 multiplies the product of the Q bit stream and a  $j$  component, received from a multiplier 910, by the spreading code  $C_{OVSF}$ . A channel compensator 905 channel-compensates the spread signals received from the multipliers 921 and 923. A summer 908 sums the channel-compensated I and Q bit streams and feeds 20 the sum to a puncturer 907.

Meanwhile, a puncturing controller 906 in the UE determines the positions inserted with an ACK/NACK relating to the uplink data transmitted on the EUDCH through an initial setup with the Node B. The puncturing controller 25 906 determines the randomly inserted positions as described in connection with Equation (2) and Equation (4). The puncturer 907 extracts the ACK/NACK from the inserted positions in the signal received from the summer 908, feeds the ACK/NACK to a buffer 905. The ACK/NACK is buffered because the Node B transmitter transmitted a  $p$ -bit ACK/NACK not in one slot at one time but 30 distributedly in  $p/N$  bits per slot for  $N$  slots of a TTI (on the assumption that one TTI has  $N$  slots). Thus, the UE receiver buffers the  $p/N$ -bit ACK/NACK extracted from each of the  $N$  slots of the TTI  $N$  times at the buffer 905, outputs the extracted  $p$ -bit ACK/NACK to an ACK/NACK extractor 904, and outputs the remaining signal as the DL DPCH signal. The ACK/NACK extractor 904 35 converts the  $p$ -bit ACK/NACK to a 1-bit ACK/NACK.

An operation for transmitting an ACK/NACK relating to the uplink data transmitted on the EUDCH will be described with reference to FIG. 10.

5 FIG. 10 is a flowchart illustrating an operation for transmitting an ACK/NACK relating to uplink data transmitted on the EUDCH according to the embodiments of the present invention.

Referring to FIG. 10, the Node B determines the number of transmission  
10 occurrences of an ACK/NACK about uplink data within one TTI through an initial setup with the UE in step 1001. Upon receipt of uplink packet data on the EUDCH, the Node B determines if the received packet data is normal in step 1002. The normal or abnormal reception is determined by a CRC check on the received packet data. If the CRC check result indicates no errors, the reception is  
15 considered normal, and if the CRC check indicates errors, the reception is considered abnormal. In step 1003, the Node B determines whether or not to transmit an ACK/NACK that relates to the uplink data according to the CRC check result.

20 The Node B generates a DL DPCH data packet to be transmitted in step 1004 and determines the positions in a data field of the DL DPCH in which the ACK/NACK is to be inserted in step 1005. The ACK/NACK positions are determined in one of the two methods expressed in Equation (1) to Equation (4). In step 1006, the Node B punctures the decided bit positions, inserts the  
25 ACK/NACK in the punctured bit positions, and transmits the DL DPCH with the ACK/NACK to the UE.

The inventive method of randomly determining the bit positions for an ACK/NACK is also applicable to other channels available in the EUDCH  
30 communication system. Also, the Node B may command the increase/decrease/maintenance of a maximum transmit power for the UE in the scheduling of step 205 shown in Fig. 2. This can be implemented by randomly puncturing a part of a DL DPCH data field similar to the random determination of the ACK/NACK positions.

In accordance with the present invention as described above, the puncturing of a data field of the existing DL DPCH and insertion of an ACK/NACK that relates to the uplink data in the punctured position in an EUDCH communication system ensures compatibility with other systems and  
5 supports HARQ for uplink data transmission.

While the invention has been shown and described with reference to certain preferred embodiments thereof, it will be understood by those skilled in the art that various changes in form and details may be made therein without  
10 departing from the spirit and scope of the invention as defined by the appended claims.